Active Magnetic Bearings for Energy Storage Systems for Combat Vehicles

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Abstract—Advanced energy storage systems for electric guns and other pulsed weapons on combat vehicles present significant challenges for rotor bearing design. Active magnetic bearings (AMB's) present one emerging bearing option with major advantages in terms of lifetime and rotational speed, and also favorably integrate into high-speed flywheel systems. The Department of Defense Combat Hybrid Power Systems (CHPS) program serves as an excellent case study for magnetic bearing applications on combat vehicles. Under the sponsorship of the CHPS program, The University of Texas at Austin Center for Electromechanics (UT-CEM) has designed active magnetic bearing actuators for use in a 5 MW flywheel alternator with a 318 kg (700 lb), 20 000 rpm rotor. The flywheel alternator serves as a power supply for multiple systems on a military vehicle, including mobility load leveling and weapons systems. Because of continuous duty requirements, magnetic bearings were chosen for this high-speed application to minimize losses and to enable the flywheel to meet a planned vehicle life of 15 to 25 years.

To minimize CHPS flywheel size and mass, a topology was chosen in which the rotating portion of the flywheel is located outside the stationary components. Accordingly, magnetic bearing actuators are required which share this "inside-out" configuration. Because of inherent low loss and nearly linear force characteristics, UT-CEM has designed and analyzed permanent magnet bias bearing actuators for this application. To verify actuator performance, a nonrotating bearing test fixture was designed and built which permits measurement of static and dynamic force. An active magnetic bearing (AMB) control system was designed to provide robust, efficient magnetic levitation of the CHPS rotor over a wide range of operating speeds and disturbance inputs, while minimizing the occurrence of backup bearing touchdowns. This paper discusses bearing system requirements, actuator and controller design, and predicted performance; it also compares theoretical vs. measured actuator characteristics.

I. INTRODUCTION

THE COMBAT Hybrid Power Systems (CHPS) program of the Department of Defense seeks to develop a new class of hybrid-electric combat vehicles with significant advances in the areas of mobility, armaments, and defensive systems. One of the distinguishing features of these vehicles is that all major

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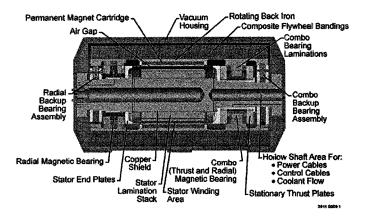


Fig. 1. CHPS flywheel alternator.

subsystems will be electrically driven, including a hybrid electric drive-train system, electric armaments, and electromagnetic armor.

The University of Texas at Austin Center for Electromechanics (UT-CEM) is developing a flywheel alternator for this project to deliver pulsed power for weapons systems as well as load-leveling power for drive-train components. High power electrical loads for the alternator include support for an electrothermal-chemical (ETC) gun, high-power lasers, and electromagnetic armor. The alternator also powers ac induction traction motors used for propulsion and a fully active electromagnetic suspension system. Energy storage for this machine is 25 MJ at 20 000 rpm with an estimated mass of 595 kg (1310 lb). It provides up to 5 MW peak power for intermittent weapons system loads and 350 kW continuous power for propulsion-related loads [1], [2].

A cross-sectional view of the CHPS flywheel alternator is shown in Fig. 1; note that the alternator axis is oriented vertically in the vehicle. This machine consists of a vacuum housing that supports a hollow, nonrotating shaft (stator), and a magnetically levitated composite rotor. The inside-out topology shown is very efficient in terms of maximizing energy stored per unit volume, but creates challenging bearing requirements. As magnetic bearings appear to be the only type compatible with the requirements for the CHPS flywheel, development of inside-out bearing actuators and a robust control system is essential to successful flywheel design.

II. BEARING SYSTEM REQUIREMENTS

In addition to dealing with rotor static and dynamic loads common to all high-speed machines, the bearing system must

TABLE I				
MAGNETIC BEARING ACTUATOR DESIGN GOALS				

Maximum Speed	20,000 rpm
Normal Operating Speed Range	15,000 20,000 rpm
Load Capacity	2g dynamic load capacity (plus 1g static load)
Maximum Power Loss	~100 W per bearing
Frequency Response Radial Actuators	Bandwidth at least 10X machine rotational frequency [2] Full load capacity at 2X machine rotational frequency
Frequency Response Thrust Actuator	Bandwidth of approximately 10 Hz desired
Backup Bearing Requirements	Provide support when magnetic bearings not levitated Shut down from full speed with magnetic bearing system inactive Provide support for short duration loads exceeding magnetic bearing load capacity (up to 8g maximum)

accommodate the terrain loads encountered by a combat vehicle over off-road terrain. To reduce windage power losses, the alternator rotor operates in a vacuum, which demands that the bearing system be vacuum-compatible. Cooling oil is available in the stator shaft at 70 to 90°C. A summary of magnetic bearing actuator design goals appears in Table I.

III. BEARING SYSTEM DESIGN

A. Magnetic Bearing Type

Permanent magnet bias homopolar (PMBH) magnetic bearings were selected for this application based on studies showing reduced power requirements and lower losses compared to heteropolar bearings [3]. An additional benefit of PMBH bearings is simplified control because of nearly linear current stiffness and positional stiffness characteristics.

Although PMBH bearings have been described in the literature previously [4], the CHPS machine presents an unusual application due to its inside-out topology. Since the rotational portion of the bearings is positioned outboard of the stator at a relatively large radius, bearing rotor laminations must withstand higher stresses than in conventional magnetic bearings. In addition, the magnetic air gap grows larger with increasing speed, instead of decreasing as occurs in bearings of conventional configuration. These factors complicate the bearing actuator design, and for our application, necessitate the use of unconventional bearing materials.

B. Bearing System Configuration

As shown in Fig. 1, the bearing configuration chosen consists of a radial magnetic bearing used on one end of the rotor and a combination bearing on the other. The radial bearing provides support in the radial direction only, while the combination bearing supports loads in both radial and axial directions.

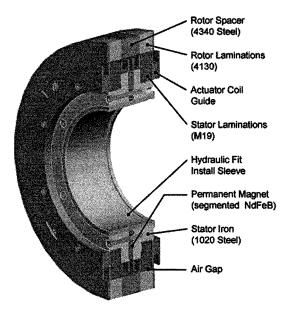


Fig. 2. Radial bearing actuator.

Also incorporated in the magnetic bearing system are backup bearings. These bearings provide support during nonoperational periods when the magnetic bearings are inactive, and provide a means to shut down the machine safely in the event of a magnetic bearing system failure. In addition, the CHPS backup bearings must occasionally provide additional support for terrain shock loads (up to 8 g maximum) which exceed the magnetic bearing load capacity.

IV. BEARING ACTUATOR DESIGN

The radial bearing actuator (Fig. 2) has an outside diameter of 36.0 cm (14.17 in.), an overall length of 11.0 cm (4.33 in.), and a radial air gap of 0.051 cm (0.020 in.) at rest. At 20 000 rpm, the radial air gap increases to 0.125 cm (0.049 in.).

In this bearing, the bias field is established with neodymium–iron–boron (NdFeB) permanent magnets located in the center of the stator. Stator laminations are fabricated from 0.356 mm (0.014 in.) thick M-19 silicon steel and include dovetail-shaped coil slots to reduce rotor lamination rotating losses. A hydraulic fit is used for bearing stator assembly to the alternator shaft for both the radial and combination bearings. Silicon steel does not have adequate mechanical properties to withstand induced rotor stresses and corresponding fatigue conditions; therefore, a structural grade steel (AISI 4130) was chosen for the rotor laminations for desirable strength and fatigue characteristics, despite less desirable electromagnetic properties.

The CHPS combination bearing actuator is shown in Fig. 3. This bearing is 30.5 cm (12.0 in.) outside diameter with an overall length of 16.7 cm (6.58 in.), and radial and thrust air gaps of 0.051 cm (0.020 in.) at rest. At 20 000 rpm, the radial air gap increases to 0.091 cm (0.036 in.).

V. ACTUATOR ANALYSIS AND PREDICTED PERFORMANCE

For actuator electromagnetic (EM) analysis, a 1-D nonlinear magnetic circuit code for inside—out actuators was developed at

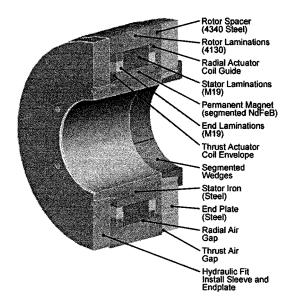


Fig. 3. Combination bearing actuator.

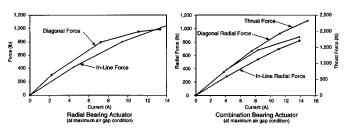


Fig. 4. Predicted actuator force versus dc current.

UT-CEM. It includes modeling of nonlinear B-H characteristics and calculations necessary for sizing of coil conductors to meet required criteria of resistance, voltage, and current. As a final check of the EM analysis, 3-D nonlinear finite element analysis (FEA) was performed.

Predicted bearing force versus dc current is shown in Fig. 4 for the radial and combination bearings operating at 20 000 rpm. In-line and diagonal radial force curves are provided, where in-line forces are defined as being in the direction of the bearing poles, and diagonal forces in the direction of the coil slots (at 45° with respect to bearing poles). To simplify control, the actuators were designed to meet stated load capacity goals while operating on the linear part of the force versus current curve.

VI. ROTORDYNAMIC ANALYSIS

To model rotordynamic behavior, an FEA model was developed at UT-CEM that enabled successful development of the controller. The model provides a complete analysis of natural frequencies and mode shapes in the range of 0 to 3000 Hz (the controller bandwidth), and includes representation of the rotor, stator shaft, end plates, and vacuum housing.

FEA results are summarized in Table II; natural frequencies and corresponding mode shapes are listed. The four lowest modes occur at frequencies below the minimum operating speed of 15 000 rpm, and represent rigid body modes involving no appreciable bending of either the rotor or stator. These modes must be traversed when the rotor accelerates to its operating speed range of 15 000 to 20 000 rpm. The fact that

TABLE II ROTORDYNAMIC FEA RESULTS

Mode Number	Frequency (cpm)	Mode Shape Description
1	62	Machine bounce mode
2	2,292	Machine conical mode
3	8,793	Rotor bounce mode
4	12,320	Rotor conical mode
5	32,980	Stator bending mode

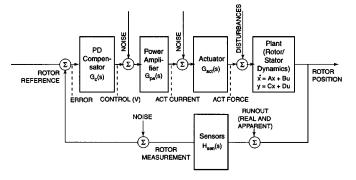


Fig. 5. Linear model of baseline PD control system.

no bending of the stator or rotor is involved means these modes can be controlled easily by the magnetic bearings.

The first flexible mode of vibration (the fundamental stator bending mode) occurs at 33 000 cpm, well above the maximum operating speed of 20 000 rpm, and will not interfere with the AMB controller as it maintains support of the rotor in its normal operating range.

VII. BEARING CONTROLLER DEVELOPMENT

Bearing controller development focused on single-input, single-output (SISO) approaches; accordingly, a proportional+derivative (PD) compensator was chosen as the baseline controller.

A linear stability analysis based on the control model shown in Fig. 5 was conducted for a range of proportional gains ($0 \le Kp \le 80$) and derivative gains ($0 \le Kd \le 0.05$). Based on this analysis, PD controller gains were selected and controller performance was evaluated for a variety of operating conditions. Using a proportional gain of 10 and a derivative gain of 0.03, simulations confirmed AMB stability over the entire speed range (0 to $20\,000\,\mathrm{rpm}$). This PD controller was robust to modest values of mass imbalance and sensor runout, but the required AMB coil voltages were excessive, and the controller was particularly sensitive to sensor noise and disturbances.

To address these deficiencies, a controller refinement called "imbalance compensation" was implemented [5], [6]. Imbalance compensation introduces a cascaded band-reject notch filter (whose center frequency tracks the rotor speed) to attenuate controller responses to synchronous disturbances. In other words, mass imbalance and synchronous runout are not controlled, which enables the rotor to spin about its inertial axis (instead of controlling it to spin about is geometric axis). Assuming adequate air gap exists, the rotor no longer transmits imbalance forces to the stator/housing (vibration and noise are reduced) and synchronous control effort is virtually eliminated.

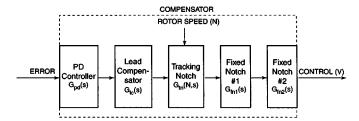


Fig. 6. Final cascaded bearing controller.

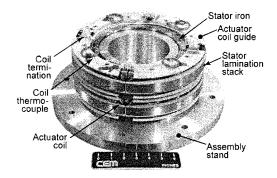


Fig. 7. Radial test bearing stator.

Unfortunately, the introduction of a tracking notch filter had an undesirable impact on overall system stability. To counteract this effect, a series of fixed-frequency notch filters and a lead compensator were systematically cascaded with the existing controller, as shown in Fig. 6. The resulting cascaded controller demonstrated acceptable performance over the operating range for a larger range of disturbances and noise.

VIII. CONTROL SYSTEM PERFORMANCE EVALUATION

An extensive series of dynamic simulations was conducted to assess system performance for a variety of operating conditions and disturbance inputs. The simulations confirmed the control system effectiveness with respect to design objectives. Most importantly, the cascaded design maintained stable levitation of the CHPS rotor over the entire speed range (0 to 20 000 rpm) for a variety of disturbance inputs, including sensor runout, mass imbalance, sensor noise, and inertial loading (due to vehicle cornering and braking). Additionally, peak coil voltages and currents were within the specified limits for the selected power amplifiers, and adequate margin against saturation in the flux path circuit was achieved. Finally, these simulations provided valuable design specifications for the power amplifiers, sensors, and machining tolerances.

IX. BEARING ACTUATOR TEST HARDWARE

UT-CEM fabricated and assembled a radial bearing actuator for testing in order to verify design and analysis codes and to provide first-hand experience with fabrication and operation of the inside—out bearing actuators. The radial test bearing stator appears in Fig. 7. To measure bearing characteristics, a nonrotating test fixture was designed and built which permitted direct measurement of output force, displacement, coil temperature, and air gap flux density levels. An illustration of the radial bearing test fixture is shown in Fig. 8. For measuring bearing

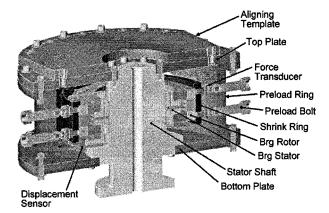


Fig. 8. Radial bearing test fixture.

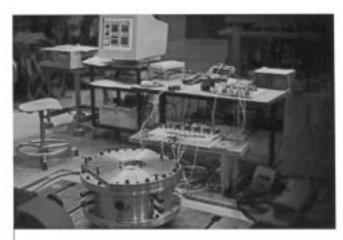




Fig. 9. Bearing test experimental hardware.

output forces, eight strain-gauge type force transducers were provided; for measuring displacements of the bearing rotor with respect to the stator, four eddy current type displacement sensors were used. Thin Hall sensors were procured that allow direct measurement of magnetic air gap flux densities. The experimental setup is shown in Fig. 9.

X. TEST RESULTS

Radial bearing actuator parameters were measured with gaps corresponding to zero-speed and full-speed conditions at room temperature. (Two rotor assemblies with different air gap dimensions were provided.) Results are summarized below.

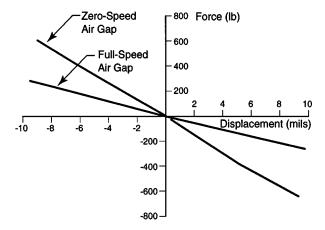


Fig. 10. Measured positional stiffness.

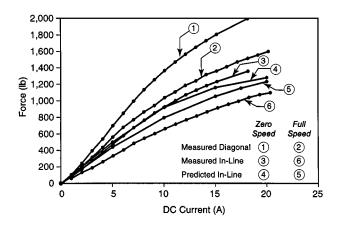


Fig. 11. Actuator force versus dc current, predicted versus measured.

A. Positional Stiffness Measurement

Predicted versus measured positional stiffness is shown in Fig. 10. Measured stiffness for the zero-speed gap is 123 kN/cm (70 lb/mil), compared to 63.1 kN/cm (36 lb/mil) for the full-speed gap. As expected, this actuator exhibits highly linear positional stiffness.

B. Output Force Measurement

Output force as a function of dc and ac actuator current is plotted in Figs. 11 and 12. Bearing design and initial force predictions were based on a lamination packing factor (PF) of 93% (the PF predicted by the lamination vendor); however, FEA results match measured results better if a 98% PF is assumed. Further study is required, therefore, to determine the actual PF and to determine the reason why 98% PF must be used in the FEA code to produce good agreement with measured results.

In Fig. 12, force versus current is plotted for ac current at 50 Hz. As seen by comparing Figs. 11 and 12, the current stiffness (force per unit current) is significantly greater for dc than for ac current input. This characteristic (which is predicted by the FEA code) is also seen in the transfer function plot of Fig. 14. These force measurements demonstrate that the actuator meets stated load capacity requirements and exhibits nearly linear behavior.

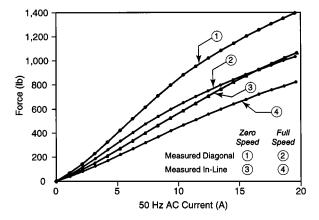


Fig. 12. Measured actuator force versus ac current.

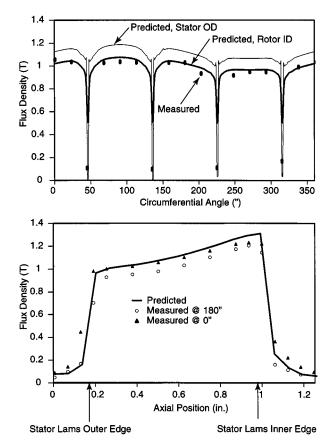


Fig. 13. Zero-speed air gap flux density distribution, predicted versus measured.

C. Air Gap Flux Density Measurement

Hall sensors were inserted into the bearing air gap to measure the bias flux density distribution. Predicted and measured field distributions as a function of circumferential angle and axial length are plotted in Fig. 13 for the zero-speed air gap condition. The predicted field levels are plotted for a 98% PF, which gives good agreement with experimental results.

D. Transfer Function Measurement

Current stiffness magnitude and phase angle versus frequency are plotted in Fig. 14 for actuator currents of 1 and 5 A. Measurements were limited to a maximum frequency of 200 Hz to

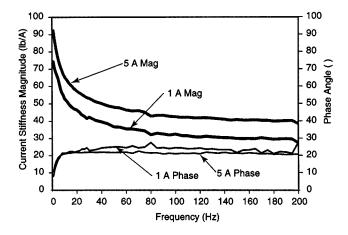


Fig. 14. Measured zero-speed actuator transfer function.

avoid bearing fixture natural frequency modes above 200 Hz. Two transfer function plot characteristics are especially noteworthy: the current stiffness magnitude at dc is substantially greater than that measured above 50 Hz, and a consistent 20° phase lag was measured between input current and output force. Knowledge of the phase lag is particularly valuable for accurate control system modeling. Eddy current and magnetic hysteresis behavior in the various bearing components are most likely responsible for the measured phase lag.

XI. CONCLUSIONS

UT-CEM has designed an active magnetic bearing system for use in a flywheel alternator for a hybrid-electric combat vehicle. The alternator and magnetic bearing actuators incorporate an inside-out topology in which rotating components are located radially outboard of the stationary components. An SISO controller was successfully designed to provide robust bearing control under simulated conditions. A prototype radial bearing was successfully fabricated, assembled, and tested in a bearing test fixture that directly measures critical bearing parameters. Knowledge gained from this work is an important step in the successful implementation of active magnetic bearings for combat vehicle energy storage systems.

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